## 900 VDC , up to 250 KHz , 20nS rising slope, Isolated Full Bridge Driver

P/N EDR83305 is rated $4.7-\mathrm{A}(\mathrm{rms}) / 4.5 \mathrm{~kW}, 30-\mathrm{A}$ (short run), $70-\mathrm{A}$ pulse $/ 60 \mathrm{~kW}$ \} P/N EDR83307 is rated at $9.2-\mathrm{A}(\mathrm{rms}) / 8.3 \mathrm{~kW}, 60-\mathrm{A}$ (short run), $120-\mathrm{A}$ pulse $/ 110 \mathrm{~kW}$ P/N EDR83308 is rated at $13-\mathrm{A}(\mathrm{rms}) / 11.5 \mathrm{~kW}, 90-\mathrm{A}$ (short run), $200-\mathrm{A}$ pulse $/ 180 \mathrm{~kW}$

Power converters, Bipolar Permanent Magnet Stepping Motors, DC Motor, Piezo-transducers, etc


Qualified for delivering kilowatts of power in ultra-precision PWM applications


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## Under management



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Piezo Drivers<br>Video Switches<br>$1 / 2$ Bridge drivers<br>Q-type high-pass filters<br>Precision F-to-V Converters<br>Soft-Landing Solenoid Drivers<br>$50 \mathrm{~Hz} / 60 \mathrm{~Hz}$ Comb Notch filters<br>H-bridge or Full-bridge Drivers<br>Super-high Power, fast Switches<br>High-power, high-speed Switches<br>Universal Analog Building Module<br>Signal Switching Separating Network<br>Sockets for relays, switches, and drivers<br>Charge-Pump Wide-Band FM detectors<br>Low-noise, High-Voltage DC/DC converters<br>DC-3phase AC resonance mode driver for $E V$<br>DC-12phase AC resonance mode driver for EV<br>Perpetual Pulse-width Discriminator, US Patent<br>$1 / 2$ and H Fuzzy Logic sockets for various relays<br>Fuzzy-Logic SPDT Relays, Switches, and $1 / 2$ Drivers<br>Fully protected, Solid-State DPST Brake, US Patent<br>Single Pole, Single Throw Relays, and Switches (SPST)<br>Power-distributing module for Motorcycles, US Patent<br>Single Pole, Double Throw Relays, and Switches (SPDT)<br>Double Pole, Single Throw Relays and Switches (DPST)<br>1-Form B, SPST-NC (normally closed) Solid State Relays<br>Charge-and-Add, Up/Down DC/DC Converters, US patent<br>1-Form B and 1-Form A DPST-NC/NO Solid State Relays<br>$\mu$-Power Controller for Magnetic Latching Valves, US Patent<br>High Voltage, Nana-Seconds Rise/Fall time, Push-Pull Drivers<br>Super-low noise preamplifiers for low and high-impedance sources<br>$\mu$-control, High-Power SPST-NC, normally closed relays, US Patent High Speed Biases Switch (HSBS-600/601) for Magnetic Resonance Spectroscopy High-Speed Biases Generator (HSBG-602) for Magnetic Resonance Spectroscopy Dynamic Disabling Switch (DDS-700/701/702) for Magnetic Resonance Spectroscopy

We are working diligently to bring new devices to the market and to meet your requests. Above is a list of the family of devices we developed and manufactured. Most of them are covered by propriety technologies, and some of them are so unique that we filed and received patents. We stock an inventory of available products that exceed several thousand in our warehouse. We keep a small number of popular devices in stock and are ready to ship immediately. Our production capacities exceed 10,000 devices per month, with two production robots programmed and working at full speed.

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Vladimir A. Shvartsman, Ph.D.
V_Shvartsman@vsholding.com

Technology for people's ideas

A large number of HS-FBI drivers are available rated: 24VDC/50A, 75VDC/42A, 150VDC/40, 1,200VDC/2A, and many others rated at various voltages/currents.

A more complete list is offered at the end of this datasheet.

## A family of High-Speed, Full-Bridge, Isolated (HS-FBI) Drivers

A new family of EDR's third generation all-voltage, optical fullbridge ( or H ) drivers is a cost effective solution for automotive and other power control applications. Devices assembled in a small 1.95 "W x 3.95 'L x 1.2 "H panel mounting enclosure, capable of delivering up to 6kW of power. Full $3,000 \mathrm{~V}$ input-output isolation allows safe interfacing directly to low-power CMOS (or TTL) logics. Essential three available controls (EN, DIR, and BR) offer design flexibilities while making it easily adaptable to a wide range of industrial solutions. An external MCU can control output functions of a HS-FBI by providing a PWM, direction, and brake signals to a load thus allowing using it in precision speedcontrol and power delivering applications. Switching frequencies up to $500-\mathrm{KHz}$ and a pulse width as short as $400-\mathrm{nS}$ makes HS-FBI driver capable of performing the finest and highly precision power management tasks. No extra heat sink is required for driving a load continuously at rated current.

HS-FBI drivers have found applications in controlling intelligent toys, robots, appliances, power tools, relays, high-speed solenoids, power converters, dc and bipolar stepping motors, TEC, and other power devices.


FIG-1 A simplified diagram of the HS-FBI driver

## Introduction

A family of opt-isolated HS-FBI drivers designed for motion control applications, though they can be used as Class-D amplifiers, controlling the amount of power delivered to a load, driving Piezo transducers and thermoelectric cooler (TEC) or Peltier devices, etc. Utilizing CMOS's advanced processing technique and modern MOSFET power devices, the drivers are able to achieve extremely low Rds. This benefit, combined with the fast switching speed, provides EE designers with a highly efficient and reliable device for use in a wide range of industrial, space, avionics, and defense applications.

High-Speed Full Bridge (HS-FBI) drivers can function with two independent power suppliers (Vpp and $\mathrm{V}-\mathrm{cc}$ ) or from a single power source. It provides complete isolation of low-power controls from an output voltage and a high pulsing current caused by a load.

The HS-FBI drivers are built with three controls: PIN\# 3 is a DIR/CS (direction), PIN\#4 is a BRK (brake), and PIN\#5 is an EN/PWM (enable/modulation). The EN/PWM input is tied to the internal reference voltage (5V) vie 10K resistor, the DIR/CS, and BRK to the PIN\# 1 (ground) via 10K. Either a mechanical switch or any semiconductor (transistor or CMOS/TTL logic) could be used to control any of the inputs. Once power is applied, the driver is enabled unless the EN/PWM input is connected to the GND. Only the BRK control works at that stage. If the enabled feedback is connected to the GND, the driver goes to a FREE-RUN state. In that state, no current flows through the load, and a motor stops rotating briefly.

## The truth table

| INPUTS |  |  | OUTPUTS |  |
| :--- | :--- | :--- | :--- | :--- |
| DIR | EN | BRK | L1 | L2 |
| $\mathbf{L}$ | $\mathbf{H}$ | $\mathbf{L}$ | OL | OH |
| $\mathbf{H}$ | H | $\mathbf{L}$ | OH | OL |
| $\mathbf{X}$ | $\mathbf{L}$ | $\mathbf{L}$ | Z | Z |
| $\mathbf{X}$ | $\mathbf{X}$ | $\mathbf{H}$ | OL | $\mathbf{O L}$ |

L = Low logic level; H = High logic level
Z = High Impedance (off-state)
$\mathrm{OH}=$ Output High (sourcing current to the output terminal)
OL = Output Low (sinking current from the output terminal)

X = Don' t Care

| OUTPUTS |  | Load (DC Motor) |
| :--- | :--- | :--- |
| L1 | L2 |  |
| $\mathbf{O L}$ | OH | Moves right (CW) |
| $\mathbf{O H}$ | OL | Moves left (CCW) |
| Z | Z | Free runs |
| OL | OL | Brakes (sudden stop) |




FIG-2. The HS-FBI driver

Two options are available for managing the output: (1) the EN/PWM is low, or (2) the BRK is high. Grounding (applying "low" onto) the EN effectively disconnects a load from Vpp. Applying "high" on the BRK shorts the output terminals. In this case, when the load is a DC Motor both command can be used for stopping the motor's rotation. The BRK stops suddenly or abruptly the motor's rotation, when the EN lets the motor 'free run' to a stop. The table on the left has summarized those

NOTE: In general, the brake control can be applied at any time, though we recommend prior to it executing the EN command for a short period. Once the BRK control applied a significant current rush flows through output transistors. The amount of current depends on the motor's speed and its mass, and the mass of a motor's load, in short the system inertia. As a rule, HS-FBI drivers were design withstanding a "rush current" at least 10x of the rated. Letting the system's friction to dispose some of that energy would be a wise solution.

## Vpp (power supply)

Bypass capacitors must be connected to power supply terminals $+\mathrm{Vpp} /$ CND physically as close as possible for preventing local parasitic oscillation and overshoots. Depending on a maximum consumption current, a highfrequency electrolytic capacitor of at least $10 \mu \mathrm{~F}$ and additional a ceramic capacitor $1.0 \mu \mathrm{~F}$ or greater value directly soldered to power pins for high frequency bypassing. It is rather difficult to calculate required values and an experimental trail would provide the best result.


FIG-3 Capacitors C6 \& C5 to "clean up" the Vpp

- 4 -


## Functions and Basic Operation

The HS-FBI drivers were designed as full bridge drivers for delivering either a steady, pulsing, or an alternative power. They made for a broad range of applications, including a bio-directional speed control for DC Motors and generating full voltage swings for driving piezo transducers. H-drives have only three control inputs (EN/PWM, DIR/CS, and Brake). And are ready to perform various jobs with just a few external components for controlling the start/stop and directions of the rotation. External components would expand applications into delivering a precise amount of power. The EN and DIR controls are high-frequency inputs. The EN can be used for PWM and the DIR for driving a load with an alternated voltage that is twice applied ( 2 xVpp ). Since it's capable of delivering pulses with a resolution of $400-\mathrm{nS}$, the HS-FBI driver is best suited for maintaining DC Motor speed/torque precisely and providing an exact amount of power to any other type of load.

NOTE: Once both powers (Vpp and Vcc) are applied and controls (EN/PWM and BR ) are left unconnected, the power will be presented on the output terminals (L1 and L2).


FIG-4. The highest alternated frequency is limited in part by the duration of "dead" time. As shown on the recording, the 'dead" time is equal to $485-\mathrm{nS}$. For a high-performance switch, $485-\mathrm{nS}$ long "dead" time is excessive, and it was made just for a demonstration. It's usually set to $200-\mathrm{nS}$. It is a very efficient device that has very high-speed rising/falling slopes (less than 20nS).

Clockwise (CW) and Counterclockwise (CCW) rotations: A required direction of rotation is easy to select by applying a proper voltage onto the DIR input (PIN\#3), leaving it unconnected. For a small DC Motor where a sudden stop would cause no damage, there are no requirements for stopping the motor prior to changing the direction of its rotation. The DIR can be changed at any time. A heavy-loaded and powerful motor requires some time to stop until the rotation is completed and can be reversed.

Free continue rotation: As mentioned above, applied power to a load (DC motor) can be interrupted and resumed at any time. That is easily accomplished by connecting the EN (PIN\# 5) to the CND (PIN\# 1) and rereleasing it. Since the EN is connected to +5 VDC via a 10 K resistor, any switch, relay, CMOS, TTL, or transistor could remove power from a load by connecting the EN to the ground.

If a motor turns in the wrong direction, re-connect the wires on the motor or output terminals (L1 and L2).
Braking (stop): The HS-FBI output terminals (L1 and L2) can be shortened by applying +5 VDC into PIN\# 3. The BRAKE signal overrides any other commands. As long as the brake signal is high, the output terminals remain short. When the BRAKE control is high, the gates of both output transistors Q1 and Q3 are driven to high (FIG.1) and, at the same time, block voltages to gates of Q2 and Q4 MOSFETs.

Care should be taken to ensure that the maximum ratings of the device are not exceeded in worse-case braking situations - high-speed and high-inertia loads.


FIG-5. The top is a control signal, and below are waveforms on a load at 250.9 KHz . With 80 V voltage applied, the voltage on a load is $160 \mathrm{~V} \mathbf{p - p}$. A control voltage scale is $1: 1$; the output scale is 1:100.

Driving a load with alternative power: The direction (DIR) input is designed to change the current flow through a load, thus helping to command a direction of DC Motor rotations, which is not performed frequently. The ability to drive high-speed solenoids has undoubtedly expanded the usefulness of the DIR input, but there is more. The DIR is a high-speed (frequency) input that allows one to "change directions" of the current at extremely high frequencies, FIG.5. That expends applications in the field of DC/DC converging and especially makes the HS-FBI driver useful for low-voltage power sources, like a fusion and photovoltaic cells.


FIG-6. Applying pulses of various durations (top recording) on the EN/PWM input made the HSFBI deliver an output power accordingly

Pulse Width Modulation (PWM): The enable (EN) input was designed to disconnect a load from the V pp (power supply). In reality, it manifests as applying power onto a load only during the enable time. The EN input was designed to have a high-speed property. It's helpful in regulating an average load current by accepting an extremely long pulse to as short as a $400-\mathrm{nS}$ pulse. Such flexibility allows controlling an average output current to be maintained with a high degree of accuracy.

## Typical application

## DC Motor Speed Control with an over-current protection



## FIG-7 A simplified diagram of the HS-FI driver with overcorrect protection

DC brush motors are increasingly required for a broad range of applications, including robotics, sporting equipment, portable electronics, appliances, medical devices, automotive applications, power tools, and many other automotive fields. The motor itself is a preferred alternative because it is simple, reliable, and low-cost. An advanced and robust H -bridge driver is essential for components for controlling the motor's direction, speed, and braking - the EDR's HS-FBI drivers are designed to do just that. With the addition of a few external components, the HS-FBI becomes a DC motor controller to maintain its precise speed and provide protection from excessive current.

The above diagram demonstrates how to employ two controls. Speed control is accomplished with an IC chip (LM555) with a few peripheral components. It generates a train of pulses of various widths for delivering power onto a load (DC Motor). A more extended pulse width translates into a higher power on a load, and that turns into a higher RPM. A pulse width and, eventually, the speed of the motor are controlled with a single 100 K potentiometer.

A transistor Q24 controls the reset input of the LM555. When a voltage drops across the resistor R 4 due to the current flowing through the load reaching the cut-in voltage, which is usually about 650 mV for silicon NPN BJT, the transistor starts conducting. As a result, the voltage on the reset input, pin \#4, drops, and that, in turn, low voltage on the output pin \#3. Pin \#3, connected to the EN controls and low voltage on it, turns off the driver's output

## EXPLOITATION



## FIG-8. A typical connection of external components to the HS-FBI driver

The family of HS-FBI drivers was designed with effortless exploitation in mind and included minimum essential components for increasing its survivability. Three controlling inputs are low power, high speed, and well protected against industrial environment voltage spikes. The best result obtained is when controls come from semiconductors. Many electromechanical devices can be used for controlling a driver, and a simple de-bouncing circuitry is recommended in such cases-drivers designed to withstand pulsing current that is at least x10 above the rated current. From an example, P/N EDR83207 rated at 24amperes and 240 -amperes of pulsing current and more than $400-\mathrm{A}$ of surge. The ability to withstand a high current surge is advantageous when changing DC Motor's rotations and fast stopping. FIG-9 was prepared to demonstrate a current surge while a DC motor was stopping and rapidly accelerating rotation.


FIG-9. EDR's made H-drivers are capable of withstanding large current surges. The top recording is a voltage across the DC Motor, and the bottom is a current flow through the motor. The polarity of applied voltage changed from -20 V to +20 V and back to create CW and CCW rotations. When the polarity of the applied voltage is changed, a large current surge is generated, which is a combination of brake and start-up currents.

HS-FBI drivers offer two controls for stopping DC motors in a more orderly way, thus reducing what could generate a destructive current surge and potential structural damage. We recommend using the EN/OFF-BRAKE/ON sequence for stopping a motor and the DIR-BRAKE/OFF-EN/ON for changing the direction of its rotation. Indeed, in a case of emergency, the BRAKE could be applied.
http://www.vsholding.com

Technology for people's ideas

## 4.2 kW (rms) and 63 kW (pulse), Isolated, Full-Bridge Driver

H-driver module for DC motors, Solenoids, etc.

## General Description:

The EDR83305 belongs to the family of Full Bridge Drivers designed for motion control applications, driving high-speed solenoids and thermo-cooling devices, piezo-transducers, etc. Utilizing advance processing technique and modern $\mathrm{C} 3 \mathrm{M}^{\mathrm{TM}}$ MOSFET power devices, drivers achieved an extremely low-ON resistance while switching at a high speed. EDR's made device provides designers with extremely efficient and reliable devices for use in wide industrial, space, avionics and defense applications.

## EDR83305/c/cc, as a H7G900D4Vcs/Vcc

## Features:

- H-driver assembled in a panel mountable, aluminum die-
casting box http://www.hammondmfg.com/pdf/1590P1.pdf
- TTL and CMOS compatible inputs
- Deliver up to 4.7 -A rms at $25^{\circ} \mathrm{C}$ and 3.2-A at $85^{\circ} \mathrm{C}$
- Pulsed current 70-A (PEAK)
- Five different modes (forward rotation, reverse rotation, PWM, disable, and hard brake)
- Low Rds (ON) typically, 0.14 Ohm per shoulder
- Wide range of Vpp (output) voltage, from 0 V to 900 V
- No problem with under-voltage and it can operate from 0-V to Vpp
- Input connector either screw-type terminals or a http://www.molex.com/pdm_docs/sd/901361206_sd.pdf
- $\quad \mathrm{Vcc}$ (power supply) $=5 \mathrm{~V}, 12 \mathrm{~V}$ or a wide input from 6 V to 15 V
- $\quad$ Vcs (control signal) $=$ TTL/CMOS compatible, 1 mA
- Output terminals are M4 stand-offs (screws provided)



## Applications for H-drivers:

- DC and Stepper Motor
- Bi-directional, high-speed solenoid
- Position and Velocity servomechanisms
- Factory and hobby robots
- Numerically controlled machinery
- Piezo-transducers / Doubling applied voltage
- Directly interfaced to a low power CPU
- In any application where a load (motor) and its power supply must be isolated from a control circuitry
- Low-noise (EMI) design allows it be located nearby to sensitive equipment
- It can be use for a precise and high-frequency PWM applications
- Push-Pull (bidirectional) control for electrohydraulic valves
- Thermoelectric cooler elements (TCE)


## Pin Functions

Pin

## Functional Description

- +Vpp
- L
- R
- -V pp/GND
- +5VDC OUT
- EN/PWM

Supply voltage, up to $1,700 \mathrm{~V}$ available Output terminal Output terminal Supply voltage (return)
A low power (less than 50 mA ) source it is normally high input via 10 K pull-up resistor to $\mathbf{+ 5 V D C}$. It is a CMOS compatible, high-speed input and can be used for PWM.

- BRAKE brake input is normally low via 10K to the GND. Applying +5VDC enable both output transistors of low shoulders H-bridge to conduct simultaneously thus shorting a load.
- DIR/CS
it is normally high input via 10 K pull-up resistor to $+\mathbf{5 V D C}$. It is a CMOS compatible, high-speed input.
- +Vcc

Power Supply (12VDC or 5VDC) for the internal logic

- -Vcc/GND

Return of the Vcc

Absolute Maximum Ratings for EDR83305/2/3 or D7G900D4/5/12

|  | Parameter | Max. | Units |
| :--- | :--- | :---: | :---: |
| Vpp | Power Supply | 900 | V |
| $\mathrm{Id} @ \mathrm{Tc}=25^{\circ} \mathrm{C}$ | Continuous current (average) | 4.7 | A |
| $\mathrm{Id} @ \mathrm{Tc}=85^{\circ} \mathrm{C}$ | Continuous current (average) | 3.2 | A |
| $\mathrm{Idm} \mathrm{@} \mathrm{Tc}=25^{\circ} \mathrm{C}$ | Maximum continues current $/ .1 \mathrm{sec}$ | 20 | A |
| Pd@ $\mathrm{Tc}=25^{\circ} \mathrm{C}$ | Power Dissipation at 4.7-A current | 3.75 | W |
| Pd@ $\mathrm{Tc}=85^{\circ} \mathrm{C}$ | Power Dissipation at 3.2-A current | 1.8 | W |
| Ids @ $\mathrm{Tc}=25^{\circ} \mathrm{C}$ | Surge Current | 70 | A |
| Vcc | Power Supply to the internal logic | 12 | V |
| Top | Operating temperature | -40 to 90 | ${ }^{\circ} \mathrm{C}$ |
| Tstg | Storage Temperature | -55 to 135 | ${ }^{\circ} \mathrm{C}$ |

Electrical Characteristics @ $\mathbf{T j}=25^{\circ} \mathrm{C}$ (unless otherwise specified), Vcc $=\mathbf{1 2 V}$, $\mathrm{Vpp}=\mathbf{8 0 0 V}$

|  | Parameters | Min. | Typ. | Max | Units | Conditions |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- |
|  | INPUT |  |  |  |  |  |
| Vcc | Supply voltage | 11 | 12 | 13 | VDC |  |
| Icc | Supply current, Vcc $=12 \mathrm{~V} / 10 \mathrm{KHz}$ |  | 40 |  | mA | Up to 10 KHz |
| Icc | Supply current, Vcc $=12 \mathrm{~V} / 100 \mathrm{KHz}$ |  | 100 |  | mA | Above 100 KHz |
| Via | High-level input voltage (EN) | 3.2 | 5 | 7 | V | Connected via 10K to +5VDC |
| Vil | Low-level input voltage (EN) | 0.9 | 1.0 | 1.2 | V |  |
| Vbrf | Brake control (BR) OFF | 0 | 0 | .5 | V | Connected via 10K to GND |
| Vbron | Brake control (BR) ON | 4.2 | 5 | 6 | V | +5 V ref. recommended |
| Vinl | Direction (DIR) |  |  | 0.9 |  | Low-level input voltage |
| Vinh | Direction (DIR) | 3.15 |  |  |  | High-level input voltage |
| Iinc | Input current to any control |  |  | 1.0 | mA |  |
|  | OUTPUT: Load is 300 Ohm |  |  |  |  |  |
| Vpp | Supply | 0 | 700 | 800 | V |  |
| Icc | Output Disable | 0.16 | 0.17 | 0.18 | Ohm | In either direction, CW \& CCW |
| Rds | Output Total resistance |  | 1 | 100.0 | $\mu \mathrm{~A}$ | Vpp=900V |
| Ill | Output leakage current |  | 10 |  | nS |  |
| T r-slope | Rising slope |  | 300 | 310 | nS |  |
| Tplh | Propagation delay turn-on time |  | 300 | 345 | nS |  |
| Tphl | Propagation delay turn-off time |  |  | 40 | nS |  |
| Trev | Propagation delay, phase reverse |  |  | 1500 |  | nS |
| Tdtm | "Dead" time |  |  | 600 | nS | Load resistive |
| P | Pulse width (minimum) |  |  | 250 | KHz | Load resistive |
| F | Maximum switching frequency |  |  |  |  |  |

## PIN FUNCTIONS (refer to the block diagram)

| PIN \# | NAME | FUNCTION |
| :---: | :---: | :--- |
| 10 | - Vpp | Power Supply Return for the Output Stage (Vpp) ground |
| 9 | L1 | Output L1 of the Bridge, the current flows through the load connected between <br> and the second output L2. |
| 8 | L2 | Output L1 of the Bridge, the current flows through the load connected between <br> and the second output L2. |
| 7 | + Vpp | Supply Voltage for the Power Output Stage. A non-inductive <1.0mF capacitor <br> must be connected between this pin and -Vpp/GND. |
| 6 | +5 VDC | +5Vref out., 20mA max |
| 5 | EN | CMOS/TTL Compatible input of the bridge to enable/disable outputs and turn the <br> driver into a stand-by state |
| 4 | BRAKE | CMOS/TTL Compatible input of shorting the load |
| 3 | DIR | CMOS/TTL Compatible input of the bridge to set a direction of rotation |
| 2 | +Vcc | Supply Voltage for the internal logic. |
| 1 | GND | Return of the Vcc. |

Absolute Maximum Ratings for EDR83307/2/3 or H7G900D10/5/12 Vcc $($ power supply $)=5 \mathrm{VDC}$ and Vcs $($ control signal $)=12 \mathrm{~V}$

|  | Parameter | Max. | Units |
| :--- | :--- | :---: | :---: |
| Vpp | Power Supply | 900 | V |
| $\mathrm{Id} @ \mathrm{Tc}=25^{\circ} \mathrm{C}$ | Continuous current (average) | 9.2 | A |
| $\mathrm{Id} @ \mathrm{Tc}=85^{\circ} \mathrm{C}$ | Continuous current (average) | 6.1 | A |
| $\mathrm{Idm} \mathrm{@} \mathrm{Tc}=25^{\circ} \mathrm{C}$ | Maximum continues current $/ .1 \mathrm{sec}$ | 40 | A |
| Pd@ $\mathrm{Tc}=25^{\circ} \mathrm{C}$ | Power Dissipation at 9.2-A current | 3.9 | W |
| Pd@ Tc $=85^{\circ} \mathrm{C}$ | Power Dissipation at 6.1-A current | 2.0 | W |
| Ids @ $\mathrm{Tc}=25^{\circ} \mathrm{C}$ | Surge Current | 120 | A |
| Vcc | Power Supply to the internal logic | 12 | V |
| Top | Operating temperature | -40 to 90 | ${ }^{\circ} \mathrm{C}$ |
| Tstg | Storage Temperature | -55 to 135 | ${ }^{\circ} \mathrm{C}$ |

Electrical Characteristics @ $\mathbf{T j}=\mathbf{2 5}^{\circ} \mathbf{C}$ (unless otherwise specified), Vcc $=\mathbf{1 2 V}, \mathbf{V p p}=\mathbf{8 0 0 V}$

|  | Parameters | Min. | Typ. | Max | Units | Conditions |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- |
|  | INPUT |  |  |  |  |  |
| Vcc | Supply voltage | 11 | 12 | 13 | VDC |  |
| Icc | Supply current, Vcc $=12 \mathrm{~V} / 10 \mathrm{KHz}$ |  | 40 |  | mA | Up to 10 KHz |
| Icc | Supply current, Vcc $=12 \mathrm{~V} / 100 \mathrm{KHz}$ |  | 100 |  | mA | Higher than 100 KHz |
| Vih | High-level input voltage (EN) | 3.2 | 5 | 7 | V | Connected via 10K to +5VDC |
| Vil | Low-level input voltage (EN) | 0.9 | 1.0 | 1.2 | V |  |
| Vbrf | Brake control (BR) OFF | 0 | 0 | .5 | V | Connected via 10K to GND |
| Vbron | Brake control (BR) ON | 4.2 | 5 | 6 | V | +5V ref. recommended |
| Vinl | Direction (DIR) |  |  | 0.9 |  | Low-level input voltage |
| Vinh | Direction (DIR) | 3.15 |  |  |  | High-level input voltage |
| Iinc | Input current to any control |  |  | 1.0 | mA |  |
|  | OUTPUT: Load is 300 Ohm |  |  |  |  |  |
| Vpp | Supply | 0 | 700 | 800 | V |  |
| Icc | Output Disable |  |  | 2 | $\mu \mathrm{~A}$ |  |
| Rds | Output Total resistance |  | 1 | 100.0 | $\mu \mathrm{~A}$ | Vpp=900V |
| Ill | Output leakage current |  | 10 |  | nS |  |
| T r-slope | Rising slope |  | 300 | 310 | nS |  |
| Tplh | Propagation delay turn-on time |  | 300 | 345 | nS |  |
| Tphl | Propagation delay turn-off time |  |  | 40 | nS |  |
| Trev | Propagation delay, phase reverse |  |  | 1500 |  | nS |
| Idtm can be set at 100nS and higher. |  |  |  |  |  |  |
| P | "Dead" time | Pulse width (minimum) |  |  | 600 | nS |
| F | Maximum switching frequency |  |  | 250 | KHz | Load resistive |

## PIN FUNCTIONS (refer to the block diagram)

| PIN \# | NAME | FUNCTION |
| :---: | :---: | :--- |
| 10 | - Vpp | Power Supply Return for the Output Stage (Vpp) ground |
| 9 | L1 | Output L1 of the Bridge, the current flows through the load connected between <br> and the second output L2. |
| 8 | L2 | Output L1 of the Bridge, the current flows through the load connected between <br> and the second output L2. |
| 7 | + Vpp | Supply Voltage for the Power Output Stage. A non-inductive <1.0mF capacitor <br> must be connected between this pin and -Vpp/GND. |
| 6 | +5VDC | +5Vref out., 20mA max |
| 5 | EN | CMOS/TTL Compatible input of the bridge to enable/disable outputs and turn the <br> driver into a stand-by state |
| 4 | BRAKE | CMOS/TTL Compatible input of shorting the load |
| 3 | DIR | CMOS/TTL Compatible input of the bridge to set a direction of rotation |
| 2 | +Vcc | Supply Voltage for the internal logic. |
| 1 | GND | Return of the Vcc. |

# Absolute Maximum Ratings for EDR83213/2 or H7G150D10/5 Vcc $($ power supply $)=5 \mathrm{VDC} \&$ Ves $($ control signal $)=5 \mathrm{~V}$ 

|  | Parameter | Max. | Units |
| :--- | :--- | :---: | :---: |
| Vpp (max) | Applied Power Supply (voltage) | 150 | V |
| $\mathrm{Id} @ \mathrm{Tc}=25^{\circ} \mathrm{C}$ | Continuous current (average) | 10.0 | A |
| $\mathrm{Id} @ \mathrm{Tc}=85^{\circ} \mathrm{C}$ | Continuous current (average) | 8.1 | A |
| $\mathrm{Idm} \mathrm{@} \mathrm{Tc}=25^{\circ} \mathrm{C}$ | Maximum continues current $/ .1 \mathrm{sec}$ | 90 | A |
| Pd@ $\mathrm{Tc}=25^{\circ} \mathrm{C}$ | Power Dissipation at 9.2-A current | 3.9 | W |
| Pd@ $\mathrm{Tc}=85^{\circ} \mathrm{C}$ | Power Dissipation at 6.1-A current | 2.0 | W |
| $\mathrm{Ids} \mathrm{@} \mathrm{Tc}=25^{\circ} \mathrm{C}$ | Surge Current | 120 | A |
| Vcc (max) | Power Supply to the internal logic | 5.4 | VDC |
| Top | Operating temperature | -40 to 90 | ${ }^{\circ} \mathrm{C}$ |
| Tstg | Storage Temperature | -55 to 135 | ${ }^{\circ} \mathrm{C}$ |

Electrical Characteristics @ $\mathbf{T j}=\mathbf{2 5}^{\circ} \mathrm{C}$ (unless otherwise specified), $\mathbf{V c c}=\mathbf{1 2 V}, \mathrm{Vpp}=\mathbf{8 0 0 V}$

|  | Parameters | Min. | Typ. | Max | Units | Conditions |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- |
|  | INPUT |  |  |  |  |  |
| Vcc | Supply voltage | 11 | 12 | 13 | VDC |  |
| Icc | Supply current, Vcc $=12 \mathrm{~V} / 10 \mathrm{KHz}$ |  | 40 |  | mA | Up to 10 KHz |
| Icc | Supply current, Vcc $=12 \mathrm{~V} / 100 \mathrm{KHz}$ |  | 100 |  | mA | Higher than 100 KHz |
| Vih | High-level input voltage (EN) | 3.2 | 5 | 7 | V | Connected via 10K to +5VDC |
| Vil | Low-level input voltage (EN) | 0.9 | 1.0 | 1.2 | V |  |
| Vbrf | Brake control (BR) OFF | 0 | 0 | .5 | V | Connected via 10K to GND |
| Vbron | Brake control (BR) ON | 4.2 | 5 | 6 | V | +5V ref. recommended |
| Vinl | Direction (DIR) |  |  | 0.9 |  | Low-level input voltage |
| Vinh | Direction (DIR) | 3.15 |  |  |  | High-level input voltage |
| Iinc | Input current to any control |  |  | 1.0 | mA |  |
|  | OUTPUT: Load is 300 Ohm |  |  |  |  |  |
| Vpp | Supply | 0 | 700 | 800 | V |  |
| Icc | Output Disable |  |  | 2 | $\mu \mathrm{~A}$ |  |
| Rds | Output Total resistance |  | 1 | 100.0 | $\mu \mathrm{~A}$ | Vpp=900V |
| Ill | Output leakage current |  | 10 |  | nS |  |
| T r-slope | Rising slope |  | 300 | 310 | nS |  |
| Tplh | Propagation delay turn-on time |  | 300 | 345 | nS |  |
| Tphl | Propagation delay turn-off time |  |  | 40 | nS |  |
| Trev | Propagation delay, phase reverse |  |  | 1500 |  | nS |
| Idtm can be set at 100nS and higher. |  |  |  |  |  |  |
| P | "Dead" time | Pulse width (minimum) |  | 600 | nS | Load resistive |
| F | Maximum switching frequency |  |  | 250 | KHz | Load resistive |

## PIN FUNCTIONS (refer to the block diagram)

| PIN \# | NAME | FUNCTION |
| :---: | :---: | :---: |
| 10 | -Vpp | Power Supply Return for the Output Stage (Vpp) ground |
| 9 | L1 | Output L1 of the Bridge, the current flows through the load connected between and the second output L2. |
| 8 | L2 | Output L1 of the Bridge, the current flows through the load connected between and the second output L2. |
| 7 | +Vpp | Supply Voltage for the Power Output Stage. A non-inductive $<1.0 \mathrm{mF}$ capacitor must be connected between this pin and -Vpp/GND. |
| 6 | +5VDC | +5 V ref out., 20 mA max |
| 5 | EN | CMOS/TTL Compatible input of the bridge to enable/disable outputs and turn the driver into a stand-by state |
| 4 | BRAKE | CMOS/TTL Compatible input of shorting the load |
| 3 | DIR | CMOS/TTL Compatible input of the bridge to set a direction of rotation |
| 2 | +Vcc | Supply Voltage for the internal logic. |
| 1 | GND | Return of the Vcc. |



FIG-11
Applied signal is $50 \mathrm{KH}, \mathrm{Vpp}=100 \mathrm{~V}$, Load $=25 \mathrm{Ohm}$


FIG-12
Similar to the recording shown in FIG-11 accept, an applied frequency was 125 KHz

## PWM, Voltage doubling, and Driving Piezo-transducers



FIG-13
All controls, such as DIR, ENBLE, and BRAKE, were built for high-speed (frequency) operations. Depending on the application, the DIR control can be used for selecting clockwise or counterclockwise rotations while driving a DC motor. It can also be used to double an input voltage. A hooking-up, as shown in FIG-13, doubles an applied voltage, or V out $=2 \times \mathrm{Vpp}$.


FIG-14
An H-driver can be used as a DC/AC or DC/DC high-power converter by driving a transformer.

The input connector is http://www.molex.com/pdm_docs/sd/901361206_sd.pdf. All power terminals are M4.


Figure 13


Figure 14

# The third generation of all-voltage full-bridge (H-bridge) drivers 

Model numbers listed below were assembled in a " 7 " size enclosure for panel mounting.

| Model Number | $V \min$ to $V_{\text {max }}$ | Id (A) cont. | I dm | p/n |
| :---: | :---: | :---: | :---: | :---: |
| H7G24D22/v/x | 0-24 VDC | 22 A | 250 | EDR83200/I/E |
| H7G24D42/v/x | 0-24 VDC | 42 A | 500 | EDR83201/I/E |
| H7G30D26/v/x | 0-30 VDC | 26 A | 300 | EDR83012/I/E |
| H7G30D14/v/x | 0-30 VDC | 14 A | 180 | EDR83009/I/E |
| H7G40D16/v/x | 0-40 VDC | 16 A | 200 | EDR83202/I/E |
| H7G40D28/v/x | $0-40$ VDC | 28A | 290 | EDR83204/I/E |
| H7G55D18/v/x | $0-55 \mathrm{VDC}$ | 18A | 200 | EDR83205/I/E |
| H7G55D24/v/x | $0-55 \mathrm{VDC}$ | 24A | 280 | EDR83206/I/E |
| H7G60D5/v/x | $0-60 \mathrm{VDC}$ | 5 A | 50 | EDR82985/I/E |
| H7G60D9/v/x | $0-60$ VDC | 8.5 A | 90 | EDR82998/I/E |
| H7G60D24/v/x | $0-60$ VDC | 24 A | 240 | EDR83207/I/E |
| H7G60D40/v/x | $0-60 \mathrm{VDC}$ | 40 A | 440 | EDR83221/I/E |
| H7G60D70/v/x | $0-60 \mathrm{VDC}$ | 70 A | 500 | EDR83304/I/E |
| H7G75D15/v/x | 0-75 VDC | 15 A | 170 | EDR83208/I/E |
| H7G75D22/v/x | 0-75 VDC | 22 A | 250 | EDR83209/I/E |
| H7G75D30/v/x | 0-75 VDC | 30 A | 350 | EDR83215/I/E |
| H7G100D10/v/x | $0-100 \mathrm{VDC}$ | 10 A | 140 | EDR83210/I/E |
| H7G100D17/v/x | $0-100 \mathrm{VDC}$ | 17 A | 210 | EDR83211/I/E |
| H7G100D30/v/x | 0-100 VDC | 30 A | 350 | EDR83212/I/E |
| H7G150D10/v/x | $0-150$ VDC | 10 A | 120 | EDR83216/I/E |
| H7G150D13/v/x | $0-150$ VDC | 13 A | 150 | EDR83217/I/E |
| H7G150D24/v/x | $0-150$ VDC | 25 A | 280 | EDR83218/I/E |
| H7G150D24/v/x | $0-150$ VDC | 24 A | 300 | EDR83219/I/E |
| H7G500D2/v/x | $0-500$ VDC | 2 A | 12 | EDR83222/I/E |
| H7G600D2/v/x | $0-600$ VDC | 2 A | 12 | EDR83223/I/E |
| H7G650D3/v/x | $0-250$ VDC | 3.2 A | 20 | EDR83306/I/E |
| H7G800D2/v/x | $0-800$ VDC | 2 A | 10 | EDR83224/I/E |
| H7G900D07/v/x | $0-900$ VDC | .7A | 7 | EDR83225/I/E |
| H7G9000D5/v/x | $0-900$ VDC | 4.7 A | 30 | EDR83305/I/E |
| H7G9000D10//v/x | $0-\mathrm{900VDC}$ | 9.3 A | 60 | EDR83307/I/E |
| H7G9000D13//v/x | $0-\mathrm{-} 900 \mathrm{VDC}$ | 13 A | 90 | EDR83308/I/E |
| H7G122D03/v/x | $0-1,200 \mathrm{VDC}$ | . 3 A | 4 | EDR83226/I/E |
| H7G172D06/v/x | $0-1,700 \mathrm{VDC}$ | . 6 A | 4 | EDR83309/I/E |
| H7G172D2/v/x | $0-1,700 \mathrm{VDC}$ | 1.3 A | 8 | EDR83310/I/E |
| H7G172D3/v/x | $0-1,700 \mathrm{VDC}$ | 2.6 A | 12 | EDR83311/I/E |

Above are just samples of drivers that were assembled in the H7G package. There are hundreds of additional drivers with various voltage/current ratings available in the same package. All drivers are built with the same control circuitry, and the difference is only in the type of output transistors (powerful MOSFETs). Do not hesitate to ask for a 30VDC/1A driver if you need one. That brings some savings due to the transistors used for assembling a $30 \mathrm{VDC} / 1 \mathrm{~A}$ driver costing less than for a $30 \mathrm{VDC} / 26 \mathrm{~A}$ driver.

Our standard, off-shelf drivers are offered with two standard Vcc $=5 \mathrm{VDC}$ and 12VDC. Please do not hesitate to ask for another VCC. In many cases, it would not add to the essential cost. The same applied to Vcs (control signals).

# Selection and Ordering Instruction for EDRs made Solid State Modules such as Relays, Switches, Breakers, $1 / 2$ and H-bridge Drivers, etc. <br> Notes: During the past ten years, the rapid development of new and additional [products gave us no choice but to expand, modify, and unify part descriptions. Below is the third modification. Our module description will be marked according to the specifications below, but p/n EDRxxxxx will stay the same for items already in circulation (already sold). 



| "X" module type |  |
| :---: | :--- |
| D | Solid-State Relay or Switch with output terminals: SPST-NO (normally open) |
| R | Solid-State Relay or Switch with output terminals: SPST-NC (normally closed) |
| W | Solid-State Relay or Switch with output terminals: DPST |
| T | Drivers, such as $1 / 2$-bridge or an SPDT relay, which can work as a $1 / 2$ driver |
| M | Driver, such as a switch with a built-in PWM controller |
| H | Full-bridge (H-bridge) Driver |
| C | Relay with built-in de-bouncing or a turn-on/off delay |
| B | Solid State Breaker and brake control modules |

"A" package dimensions

|  | $0.615^{\prime \prime} \mathrm{H} \times 1.48$ " x 0.290 " W |
| :---: | :---: |
|  | 1.75 "H x 1.80 " $\mathrm{L} \times 0.595$ "W |
|  | 1.125 "H x 1.75 " $\mathrm{L} \times 0.8$ "W |
|  | 1.15 "H x 2.0 " $\mathrm{L} \times 0.92$ " W |
|  | 1.15 "H x 2.8 " $\mathrm{L} \times 1.15$ " W |
|  | DIP24, 0.375"H x 0.925 " $\mathrm{L} \times 0.53$ " W |
|  | panel mount, .82 "H x 3.95 " x 1.96 "W |
|  | . 575 "H x 1.1 "L x 2 " W |
|  | panel mount 3 " $\mathrm{H} \times 10$ " x 8 " W |
|  | DIN type enclosure, $2.36^{\prime \prime} \mathrm{H} \times 2.36^{\prime} \mathrm{L} \times 1.5$ " W , for 35 mm DIN Rail panel mount, .8 " $\mathrm{H} \times 2.275^{\prime \prime} \mathrm{L} \times 1.75^{\prime \prime} \mathrm{W}$ |

"B" Speed - A device's ability to turn ON/OFF output terminal(s) times per second
L a low-speed relay/switch, rated DC - 200 Hz , direct driving control
A a low speed relay/switch, AC input relays
N a medium speed relay/switch, rated DC -25 KHz , direct driving control G a medium speed relay/switch, rated DC -25 KHz , low current control and power F a fast relay/switch, rated up to DC -350 KHz , low current control and power S a super-fast relay/switch, rated DC -1.4 MHz , low current control and power
U a super-fast relay/switch, rated DC -1.2 MHz , direct driving control
V Fast, High Voltage Solid-State Switches with Nanoseconds rise time
"C" Output Voltage - a maximum allowed voltage between output terminals, up to 100 kV
It must be replaced with the required voltage, and we offer the closest and highest value available.
Note: In an "AC" -relay, a voltage specified a peak-to-peak maximum voltage, and the maximum VAC could be calculated
by multiplying a maximum allowed voltage by a factor of 0.7
"F" A relay can be used to control either AC, DC, or AC/DC power
A $\quad-$ a relay/switch designed to switch/chop an $\mathrm{AC} / \mathrm{DC}$ power
D - a relay/switch designed to switch/chop a DC power
"none" - relay with a SCR or TRIAC on the output to control only AC power
"H" A maximum allowed RMS CURRENT (Ampere) without a heat sink
We can manufacture a device for any required current.
"I" Some of our products use an internal DC/DC converter to provide power to the internal electronics. Varieties voltages are available: $5 \mathrm{VDC}+/-5 \%, 12 \mathrm{VDC}+/-5 \%, 24 \mathrm{VDC}+/-5 \%$ and $48 \mathrm{VDC}+/-5 \%$. For a wider input power voltage swing, please add "W" after the voltage. For an example, 24 W is for $24 \mathrm{~V}+/-12 \mathrm{~V}$.
"E" We offer several standard control voltages: 5VDC, 12VDC, 24VDC, 48VDC, 3-20VDC, and 18-38VDC. Please specify the input control voltage, for example, D1L30D12/xx. Replace xx with a $3,5,12,24,48,3-20$, and $18-38$ that is for $3 \mathrm{VDC}, 5 \mathrm{VDC}, 12 \mathrm{VDC}, 24 \mathrm{VDC}, 48 \mathrm{VDC}, 3-$ 20 VDC and $18-38 \mathrm{VDC}$. Respectful control voltage is represented at the end of the part number in the following way, for example, EDR82653/1 and EDR82653/8. Both relays are almost the same, and the difference is only an applied control voltage, " 1 " is for 3 VDC , and " 8 " is for $18-38 \mathrm{VDC}$;

| Control Voltage | $\#$ | Control Voltage | $\#$ | Control Voltage | \# |
| :---: | :---: | :--- | :---: | :---: | :---: |
| 3 VDC | 1 | 5 VDC | 2 | 12 VDC | 3 |
| 24VDC | 4 | 48 VDC | 5 | 26 VDC | 6 |
| $3-20 \mathrm{VDC}$ | 7 | $18-38 \mathrm{VDC}$ | 8 | $90-120 \mathrm{VAC}$ | 9 |


$\frac{\text { " } \mathbf{Z} \text { " A relay/switch built with the following standard isolations }}{\text { "L" or "none" }}$| type relay in 2500 V |
| :--- |
| type relay is $3000 \mathrm{~V}, 4000 \mathrm{VDC}$ ("H4"), and 5200 ("H5") VDC. |

"T" Turn-on delays; "S" for seconds, "M" for milliseconds, "U" for microseconds, M102-100 mS turn-off delay, 102 M mS - turn-on delay

